

### COST Action TU1208

"Civil Engineering Applications of Ground Penetrating Radar"

### Cylindrical Wave Approach (CWA) for electromagnetic modelling of 2D GPR scenarios

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### **Talk Outline**

- Theoretical Method
- Numerical Solution
- Results and Applications
- Conclusions and Work in Progress



### CWA: Analytical-numerical technique for the solution of the direct two-dimensional scattering problem by a finite set of buried cylinders

### Source

- Monochromatic or pulsed plane-wave
- Line sources of current or arbitrary 2D field distribution

#### Host medium

 Linear, isotropic, dielectric, half-space or finite-thickness slab

#### **Buried Objects**

- Perfectly-conducting or dielectric circular cylinders
- Scatterers with arbitrary cross-section, simulated by a suitable configuration of circular cylinders





Lossless materials in this presentation, Nicola will explain how to deal with losses!

# Monochromatic plane-wave scattering problem by N perfectly-conducting circular cylinders buried in a dielectric half-space

(IEEE Trans. Antennas and Propagation, 53(2), 2005, pp. 719-727)

- Arbitrary radii, burial depths, and distances between the obstacles
- All the cylinder-cylinder interactions, and the multiple reflections between the cylinders and the interface, are considered
- Results in near- and far-field zones, E and H polarization





The scattering problem of a plane wave impinging on a circular-section cylinder, with infinite length and in free space, can be solved analytically with an expansion of the electromagnetic field into cylindrical waves.





The interaction of a plane wave with a flat interface separating two half-spaces can be solved analytically by using the Fresnel coefficents and the Snell's law.



 $\varepsilon_1 = \varepsilon_0 \varepsilon_{r1} \qquad \varepsilon_2 = \varepsilon_0 \varepsilon_{r2}$  $\mu_1 = \varepsilon_0 \mu_{r1} \qquad \mu_2 = \varepsilon_0 \mu_{r2}$ 

Incident field

$$\mathbf{E}^{i} = E_{h}^{i} e^{-i\boldsymbol{\beta}_{i}\cdot\mathbf{r}} \mathbf{h}_{0} + E_{v}^{i} e^{-i\boldsymbol{\beta}_{i}\cdot\mathbf{r}} \mathbf{v}_{0}$$

Reflected field

$$\mathbf{E}^{r} = \Gamma_{h} E_{h}^{i} e^{-i\boldsymbol{\beta}_{r}\cdot\mathbf{r}} \mathbf{h}_{0} + \Gamma_{v} E_{v}^{i} e^{-i\boldsymbol{\beta}_{r}\cdot\mathbf{r}} \mathbf{v}_{0}$$

Transmitted field

$$\mathbf{E}^{t} = T_{h} E_{h}^{i} e^{-i\boldsymbol{\beta}_{t} \cdot \mathbf{r}} \mathbf{h}_{0} + T_{v} E_{v}^{i} e^{-i\boldsymbol{\beta}_{t} \cdot \mathbf{r}} \mathbf{v}_{0}$$

 $\mathbf{r} = x\mathbf{x}_0 + y\mathbf{y}_0 + z\mathbf{z}_0$ 

The Fresnel coefficients give the amplitude of the reflected and transmitted plane waves.

**REFLECTION COEFFICIENTS** 

Horizontal polarization

$$\Gamma_{h} = \frac{\sqrt{(\mu_{2} / \varepsilon_{2})} \cos \theta_{i} - \sqrt{(\mu_{1} / \varepsilon_{1})} \cos \theta_{t}}{\sqrt{(\mu_{2} / \varepsilon_{2})} \cos \theta_{i} + \sqrt{(\mu_{1} / \varepsilon_{1})} \cos \theta_{t}}$$
$$\Gamma_{v} = \frac{\sqrt{(\mu_{1} / \varepsilon_{1})} \cos \theta_{i} - \sqrt{(\mu_{2} / \varepsilon_{2})} \cos \theta_{t}}{\sqrt{(\mu_{1} / \varepsilon_{1})} \cos \theta_{i} - \sqrt{(\mu_{2} / \varepsilon_{2})} \cos \theta_{t}}$$

Vertical polarization

**TRANSMISSION COEFFICIENTS** 

Horizontal polarization

$$T_h = 1 - \Gamma_h$$

·Vertical polarization

 $T_v = 1 - \Gamma_v$ 



The Snell's law gives the propagation direction of the reflected and transmitted plane waves.



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•Angle of reflection

 $\theta_r = \theta_i$ 

•Angle of transmission

$$\sqrt{\varepsilon_{r1}}\sin\theta_i = \sqrt{\varepsilon_{r2}}\sin\theta_t$$

$$\Rightarrow \theta_t = \arcsin\left(\sqrt{\frac{\varepsilon_{r1}}{\varepsilon_{r2}}}\sin\theta_i\right)$$





- Scattered field represented as a superposition of cylindrical waves
- Plane-wave spectrum to take into account the reflection and transmission of cylindrical waves by the interface

Cylindrical Function  

$$CW_{m}(\xi,\zeta) = H_{m}^{(1)}(\rho) e^{im\theta} = \frac{1}{2\pi} \int_{-\infty}^{+\infty} F_{m}(\xi,n_{\parallel}) e^{in_{\parallel}\zeta} dn_{\parallel} \qquad \text{plane-wave spectrum}$$
Reflected Cylindrical Function  

$$RW_{m}(\xi,\zeta) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} \Gamma(n_{\parallel}) F_{m}(\xi,n_{\parallel}) e^{in_{\parallel}\zeta} dn_{\parallel} \qquad \text{plane-wave reflection coefficient}$$
Transmitted Cylindrical Function

$$TW_{\mathrm{m}}(\xi,\zeta,\chi) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} T_{10}(n_{\parallel}) F_{\mathrm{m}}(-n_{1}\chi,n_{\parallel}) e^{in_{1}n_{\parallel}\zeta} e^{-i(\xi+\chi)\sqrt{1-(n_{1}n_{\parallel})^{2}}} dn_{\parallel}$$



**Scattered Field** 

$$\begin{split} V_{\mathbf{S}}(\xi,\zeta) &= V_0 \sum_{\ell=-\infty}^{+\infty} J_{\ell}(n_1\rho_{\mathrm{s}}) e^{i\ell\theta_{\mathrm{s}}} \sum_{\mathrm{q}=1}^{N} \sum_{\mathrm{m}=-\infty}^{+\infty} i^{\mathrm{m}} e^{-i\mathrm{m}\varphi_{\mathrm{t}}} \underbrace{C_{\mathrm{q}\mathrm{m}}}_{\mathrm{coefficients}} \\ &\times \left[ CW_{\mathrm{m}-\ell}(n_1\xi_{\mathrm{q}\mathrm{s}},n_1\zeta_{\mathrm{q}\mathrm{s}})(1-\delta_{\mathrm{q}\mathrm{s}}) + \frac{H_{\ell}^{(1)}(n_1\rho_{\mathrm{s}})}{J_{\ell}(n_1\rho_{\mathrm{s}})} \delta_{\mathrm{q}\mathrm{s}} \delta_{\ell\mathrm{m}} \right] \end{split}$$

### Scattered-Reflected Field

$$V_{\rm sr}(\xi,\zeta) = V_0 \sum_{\ell=-\infty}^{+\infty} J_\ell(n_1\rho_{\rm s}) e^{i\ell\theta_{\rm s}} \sum_{\rm q=1}^N \sum_{\rm m=-\infty}^{+\infty} c_{\rm qm} i^{\rm m} e^{-i{\rm m}\varphi_{\rm t}}$$

$$imes RW_{\mathrm{m+}\ell}\left[-n_{1}(\chi_{\mathrm{q}}+\chi_{\mathrm{s}}),n_{1}(\eta_{\mathrm{s}}-\eta_{\mathrm{q}})
ight]$$

**Scattered-Transmitted Field** 

$$V_{\rm st}(\xi,\zeta) = V_0 \sum_{\rm s=1}^{N} \sum_{\rm m=-\infty}^{+\infty} i^{\rm m} e^{-i{\rm m}\varphi_{\rm t}} c_{\rm sm} T W_{\rm m}(\xi - \chi_{\rm s}, \zeta - \eta_{\rm s}, \chi_{\rm s})$$

### **Boundary Conditions**

• E polarization 
$$[V_t(\xi,\zeta) + V_s(\xi,\zeta) + V_{sr}(\xi,\zeta)]_{\rho_s=\alpha_s} = 0$$

• H polarization  $\left\| \frac{\partial V_{\mathbf{t}}(\xi,\zeta)}{\partial \rho_{\mathbf{s}}} + \frac{\partial V_{\mathbf{s}}(\xi,\zeta)}{\partial \rho_{\mathbf{s}}} + \frac{\partial V_{\mathbf{sr}}(\xi,\zeta)}{\partial \rho_{\mathbf{s}}} \right\|_{\rho=0} = 0$ 

$$\begin{split} \sum_{q=1}^{N} \sum_{m=-\infty}^{+\infty} A_{pm}^{qs} c_{qm} - B_{p}^{s} &= 0 \\ s &= 1, ..., N, \ p &= 0, \pm 1, \pm 2, ..., \\ A_{pm}^{qs} &= i^{m-p} e^{-im\varphi_{t}} G_{p}(n_{1}\alpha_{s}) \left\{ CW_{m-p}(n_{1}\xi_{qs}, n_{1}\zeta_{qs})(1 - \delta_{qs}) + \frac{\delta_{qs}\delta_{mp}}{G_{p}(n_{1}\alpha_{s})} + RW_{m+p} \left[ -n_{1}(\chi_{q} + \chi_{s}), n_{1}(\eta_{s} - \eta_{q}) \right] \right\} \end{split}$$

where  $G_{\rm p}(\cdot)=J_{\rm p}(\cdot)/H_{\rm p}^{(1)}(\cdot)$  for E pol. and  $G_{\rm p}(\cdot)=J_{\rm p}'(\cdot)/{H'}_{\rm p}^{(1)}(\cdot)$  for H pol.

Monochromatic plane-wave scattering problem by N <u>dielectric</u> circular cylinders buried in a dielectric half-space

(Radio Science, 40, 2005, RS6S18)

### Field Inside the q-th Cylinder

$$V_{cq}(\xi_{q},\zeta_{q}) = V_{0} \sum_{m=-\infty}^{+\infty} i^{m} e^{-im\varphi} d_{qm} \mathcal{I}_{m}(n_{cq}\rho_{q}) e^{im\theta_{q}} \xrightarrow{\text{unknown}} coefficients$$

#### **Boundary Conditions**

$$\begin{cases} V_{t} + V_{s} + V_{sr}|_{\rho_{q} = \alpha_{q}} = V_{cq}|_{\rho_{q} = \alpha_{q}} \\ \frac{\partial}{\partial \rho_{q}} (V_{t} + V_{s} + V_{sr}) |_{\rho_{q} = \alpha_{q}} = t_{q} \frac{\partial V_{cq}}{\partial \rho_{q}} |_{\rho_{q} = \alpha_{q}} \end{cases}$$

where  $t_q = 1$  for E pol. and  $t_q = (n_1/n_{cq})^2$  for H pol.

Monochromatic plane-wave scattering problem by N perfectly-conducting and dielectric circular cylinders buried in a finite-thickness slab

(IEEE Trans. Antennas and Propagation, 57(4), 2009, pp. 1208-1217, Journal of Optical Society of America A, 27(4), 2010, 687-695)



With the introduction of more interfaces between different media, the definition of

generalized multiple-reflected cylindrical functions

#### and

multiple-reflected-transmitted cylindrical functions

relevant to multiple reflection phenomena is needed.



... in a grounded slab











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- Targets buried in a non-homogeneous soil or below asphalt layer
- Through wall scattering



(JOSA A, vol. 30(8), 2013, pp. 1632-1639)





**Pulsed plane-wave** scattering problem by N perfectly-conducting and dielectric circular cylinders buried in a dielectric half-space

(IEEE Geoscience Remote Sensing Letters, 4(4), 2007, pp. 611-615)

- Sampling of the incident-field spectrum and of the spectra of the various field terms
- Solution for any sample in the spectral domain by using the CWA
- Time-domain solution by means of the inverse transform

**Current Line** scattering problem by N perfectly-conducting and dielectric circular cylinders buried in a dielectric half-space

(Progress in Electromagnetics Research, 80, 2008, pp. 179-196)

$$V_{i}(\xi,\zeta) = -V_{0}H_{0}^{(1)}\left[n_{1}\sqrt{(\xi-\chi_{L})^{2}+(\zeta-\eta_{L})^{2}}\right]$$



### Arbitrary 2D distribution of the field

Near Surface Geophysics, 13(3), 2015, pp. 219-225

Expansion of the incident field in plane waves. CWA applied to each plane wave.

Rough surface between air and soil

Near Surface Geophysics, 11(2), 2013, pp. 177-183



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### **CWA computational issues**

### Accuracy

Depends on the truncation of the involved series (cylindrical expansions)

### Memory requirements

Low.

### **Execution time**

Increases with the number of cylinders, their size and the permittivity of the involved materials.



### Series truncation to a finite number of elements



- Convergence properties of the method
- Truncation criterion

$$M_{\rm t} = 3n_1 \alpha_{\rm max}$$

### Numerical evaluation of spectral integrals

- Infinite extension of the integration domain
- Highly oscillating behavior of the integrand
- Considerable variability of the  $F_m(\xi, n_{||})$  function

$$TW_{\rm m}(\xi,\zeta,\chi) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} T_{10}(n_{\parallel}) F_{\rm m}(-n_1\chi,n_{\parallel}) e^{in_1n_{\parallel}\zeta} e^{-i(\xi+\chi)\sqrt{1-(n_1n_{\parallel})^2}} dn_{\parallel}$$

Short computational time is desirable for inverse scattering applications

#### Integration of the evanescent spectrum

- Smaller ζ values: Laguerre-Gauss quadrature formula
- Larger ζ values: decomposition of the integration interval in subintervals of suitable length, Gauss-Legendre quadrature formula in each subinterval, ε-algorithm (convergence acceleration)

Integration of the homogeneous spectrum

- Discrimination between partially and totally reflected waves
- Development of an adaptive integration technique
  - ✓ Calculus of the local frequency oscillation rate f
  - Decomposition of the integration interval in subintervals in which f behaves monotonically; further decomposition of each subinterval, by using an adaptive procedure for the evaluation of the effective oscillation period
  - ✓ Gauss-Legendre quadrature formula in each sub-subinterval





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### Convergence properties of the CWA



 $\alpha_1 = 1, \ \alpha_2 = 3, \ \alpha_3 = 5, \ \chi_1 = 10, \ \chi_2 = 5, \ \chi_3 = 15, \ \eta_1 = -10, \ \eta_2 = 0, \ \eta_3 = 10, \ n_1 = 1.4, \ \varphi_1 = 45^\circ, \ N = 3$ 

$$M_{
m t}~=~3n_1lpha_{
m max}$$





CWA is conceived for the simulation of cylindrical targets with circular section.

But also targets with arbitrary shape can be simulated!!





The arbitrary shape of a PEC target can be approximated with a suitable set of PEC small circular-section cylinders along ist border.





The total area of the approximating small cylinders has to be equal to the area of the simulated target.



Once N is fixed, the same area rule allows to calculate the size of the small cylinders. Such rule is not the optimum, but gives good results. Configurations of cylinders of different size may also be used.









Wire-grid simulation of a square metallic border



Near field:  $\xi = -0.1$ 0.7 N =16 N = 32 0.6 N = 64--N = 128 0.5 0.4 0.3 0.2 0.1 0≌ -20 -15 -10 -5 10 20 0 5 15 ζ

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#### Same-Volume Rule ----- N=8 N=16 0.16 N=32 0.14 80 N=64 Good results are obtained when N=128 0.12 ε1 the total volume (for unit length) 0.1 0.08 V<sub>st</sub> of the N modeling scatterers is equal to the 0.06 $\psi = 45^{\circ}$ volume of the simulated object 0.04 = 1.5√2, t = 0.1√2 0.02 0 -10 -5 5 10

#### **Tilted Slab**

#### **Circular Shell** 0.4 **E**<sub>0</sub> 0.35 ε0 0.35 0.3 N=16 0.3 N=32 ε1 0.25 N=64 0.25 8 N=128 0.2 |V<sub>st</sub>| 0.2 N=256 0.15 0.15 0.1 0.1 0.05 $R_{int} = 0.8, R_{ext} = 1$ 0.05 0 -10 -5 5

#### **Rectangular-Section Scatterer**



 $n_1 = 2, n_c = 5, \chi = 2.5, \xi = 1, \varphi^i = 0, E \text{ pol.}$ 

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|V<sub>st</sub>|



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#### Re(Total scattered field) [a.u.]



- Air-dielectric-dielectric background
- Buried metallic cable

a = 5.5  cm	$\epsilon_{r1} = 3.5$
d = 60  cm	$\varepsilon_{r2} = 5$
L = 10  cm	TM pol.
$\varphi_{\rm i}=0$	





 Time-shape of the incident pulsed plane-wave:



$$-\mathbf{n}^{\mathbf{i}} \cdot \boldsymbol{\rho} = g_0 e^{-\frac{\left(\tau - \mathbf{n}^{\mathbf{i}} \cdot \boldsymbol{\rho}\right)}{2}} \left[ e^{\frac{1}{2}} \cos\left(\tau - \mathbf{n}^{\mathbf{i}} \cdot \boldsymbol{\rho}\right) - 1 \right]$$
  
Frum and
$$H(\psi, \boldsymbol{\rho}) = G(\psi) e^{i\psi(\xi \cos\varphi_{\mathbf{i}} + \zeta \sin\varphi_{\mathbf{i}})} = g_0 \frac{\sqrt{2\pi}}{\omega_0} e^{-\frac{\psi^2}{2}} \left[ \cosh\left(\psi\right) - 1 \right] e^{i\psi\mathbf{n}^{\mathbf{i}} \cdot \boldsymbol{\rho}}$$

 ✓ Waveform consistent with many current ultrawide-band short-pulsed sources

✓ This pulse can be considered vanishing for  $\tau$ >4 and its spectrum does not contain frequencies  $\psi$ >6

✓ With an appropriate choice of  $\omega_0$ the reference frequency, related to spatial resolution, can be tuned



- Time-shape of the scattered-transmitted field:
- ✓ Diffraction from cylinders
- ✓ Reverberations between cylinders and air-ground interface



# **CWA: Conclusions**

- □ CWA: Analytical-numerical method for the solution of the 2D scattering problem by a set of cylinders buried in a dielectric half-space or finite-thickness slab
  - Adaptive integration techniques for the evaluation of the spectral integrals
  - Source: monochromatic or pulsed plane-wave, current line, arbitrary 2D distribution of the field
  - Rough interfaces between different media
  - Scatterers of arbitrary cross-section simulated by a suitable set of circular cylinders
  - Results in near- and far-field zones, E and H polarization, spectral and time domain

### □ Applications:

- Characterization of GPR scenarios, discretized in cylindrical cells
- Reflection and transmission by reinforced concrete
- Diagnostics of buried utilities (tunnels, conduits, electricity cables, gas or water pipes)
- Direct solver in iterative algorithms for the solution of inverse scattering problems

### Lossy media & Extension of the method to 3D geometries:

See Nicola's presentation!



# Thank you! Join COST Action TU1208!

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